

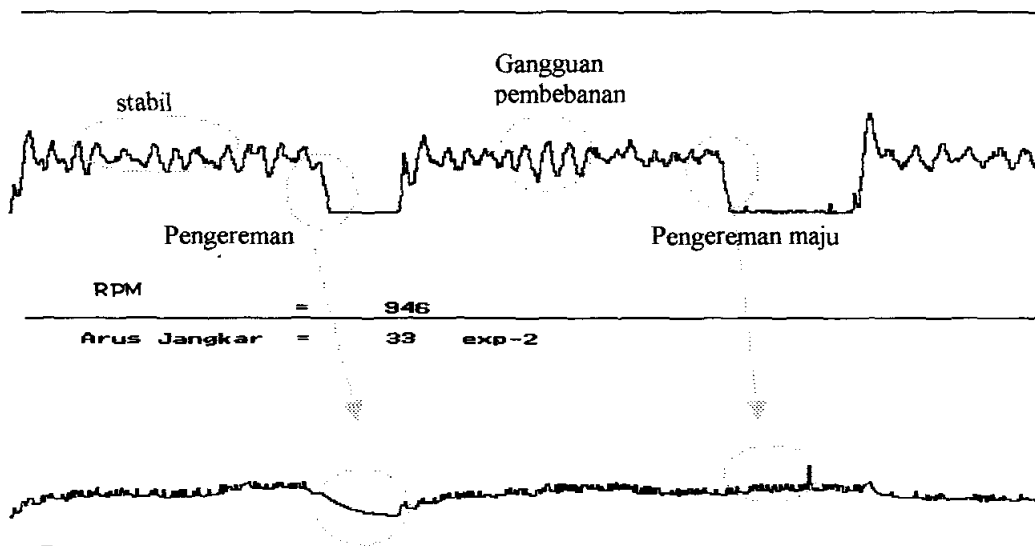


LAMPIRAN

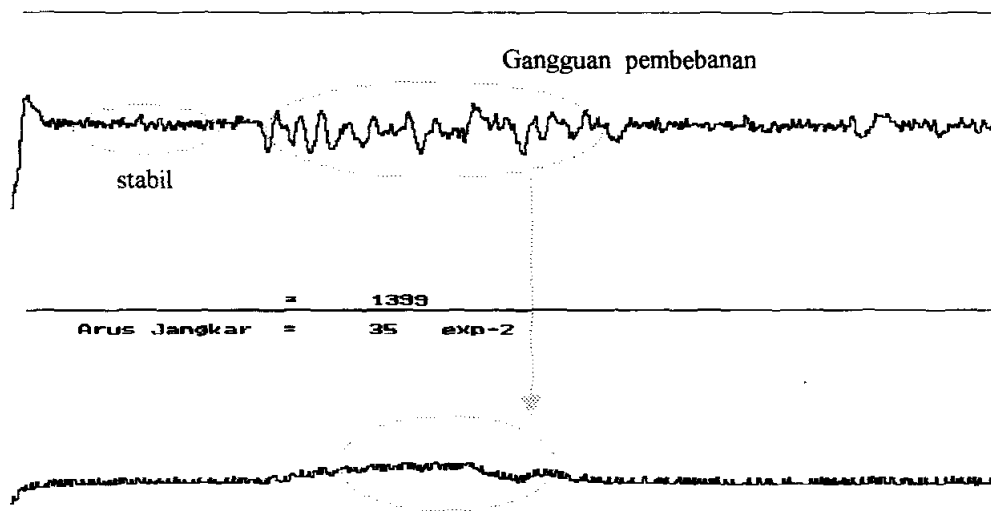
LAMPIRAN

a. Tampilan Respon Pada PC

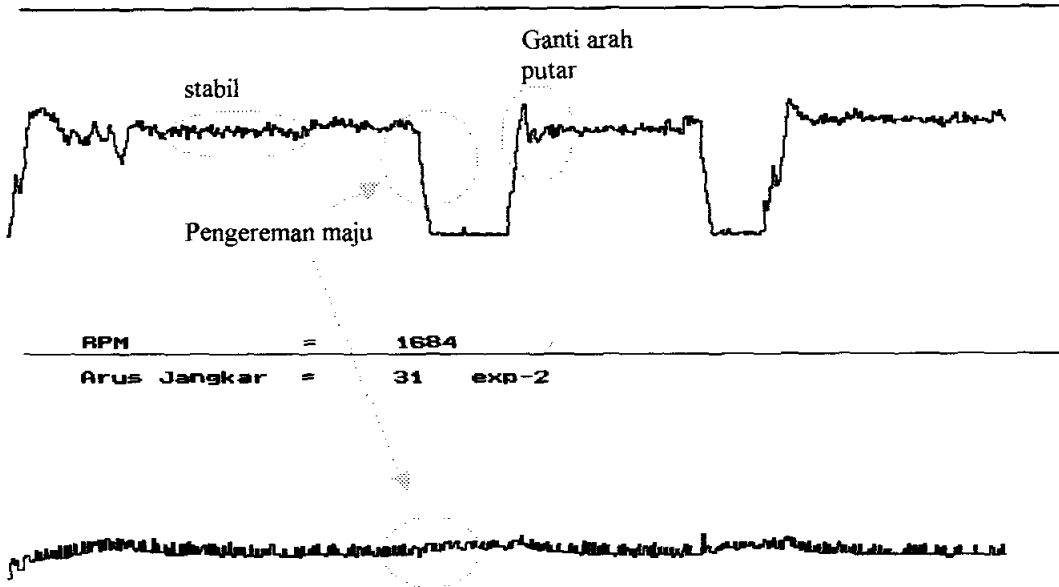
Grafik a.1. Tampilan layar Monitor PC saat motor berputar 950 RPM



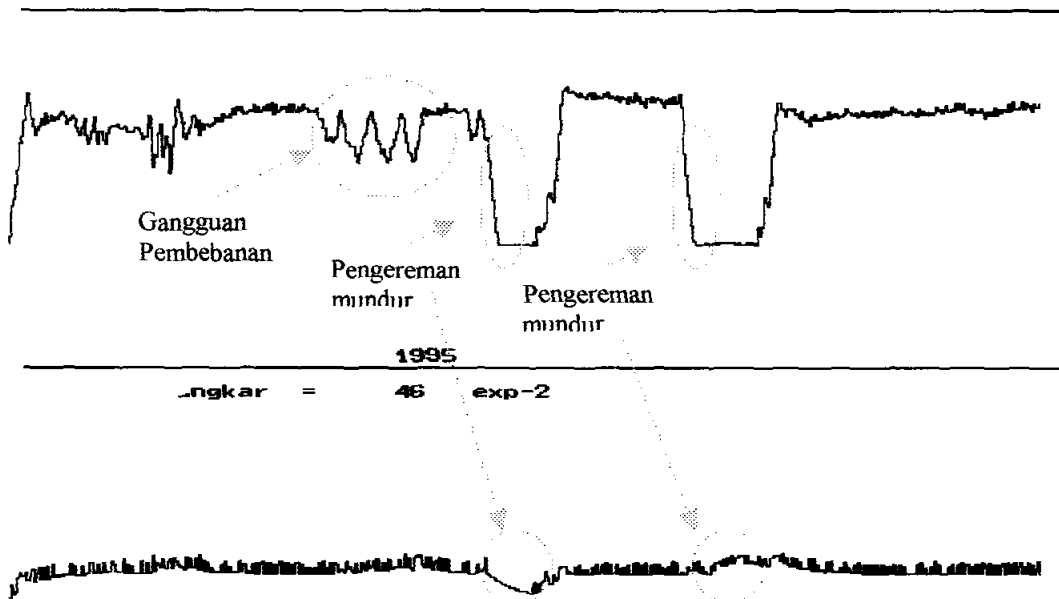
Grafik a.2. Tampilan layar Monitor PC saat motor berputar 1400 RPM



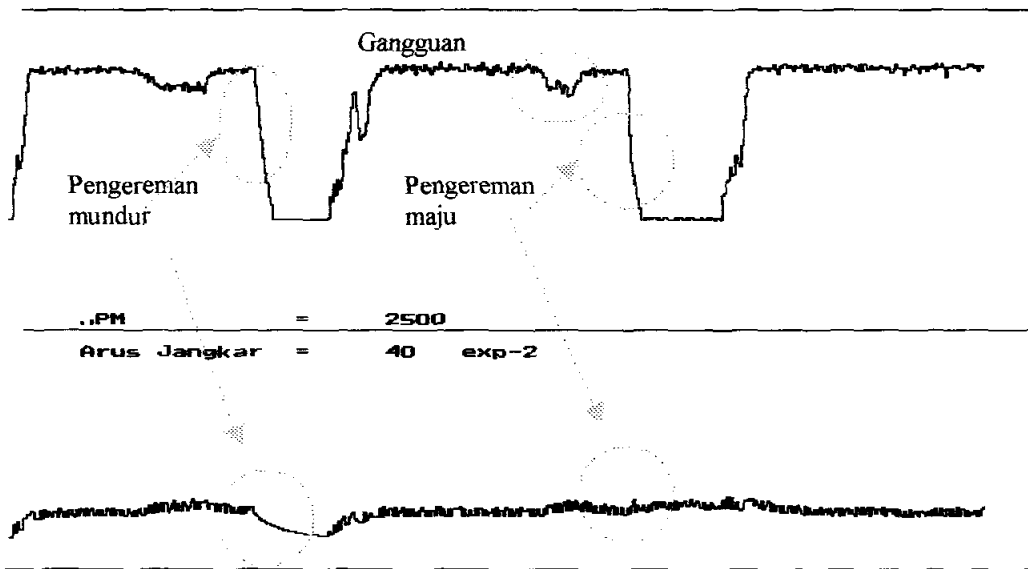
Grafik a.3. Tampilan layar Monitor PC saat motor berputar 1400 RPM



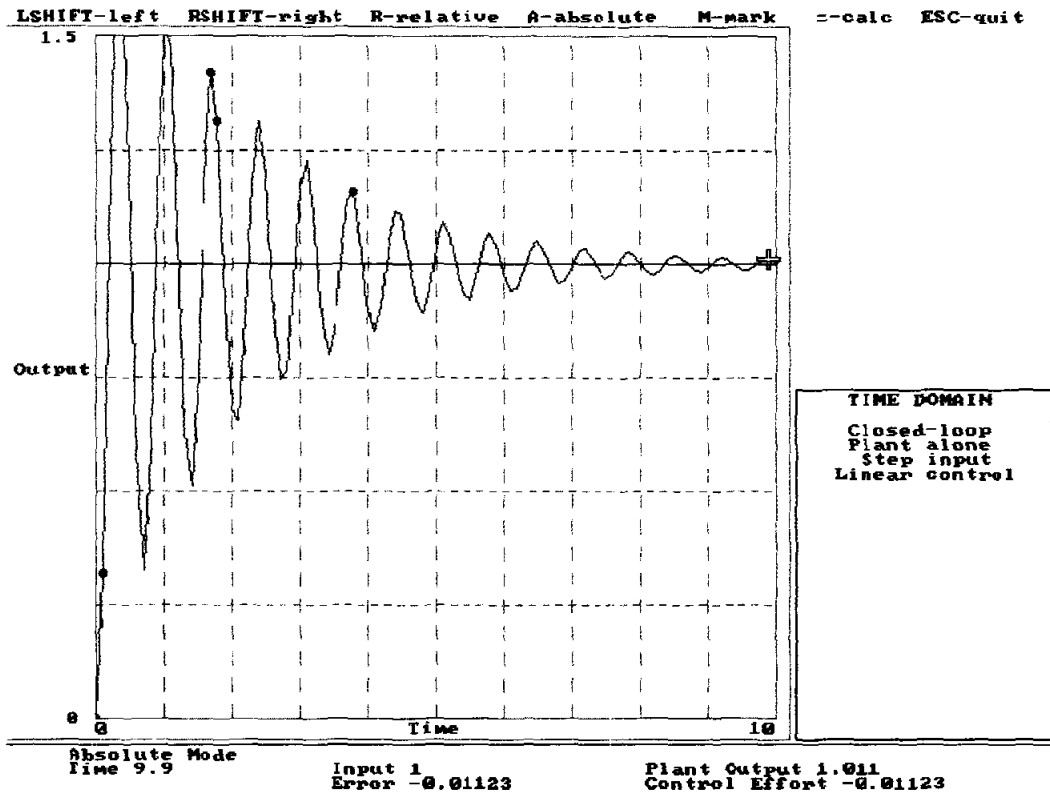
Grafik a.4. Tampilan layar Monitor PC saat motor berputar 2000 RPM



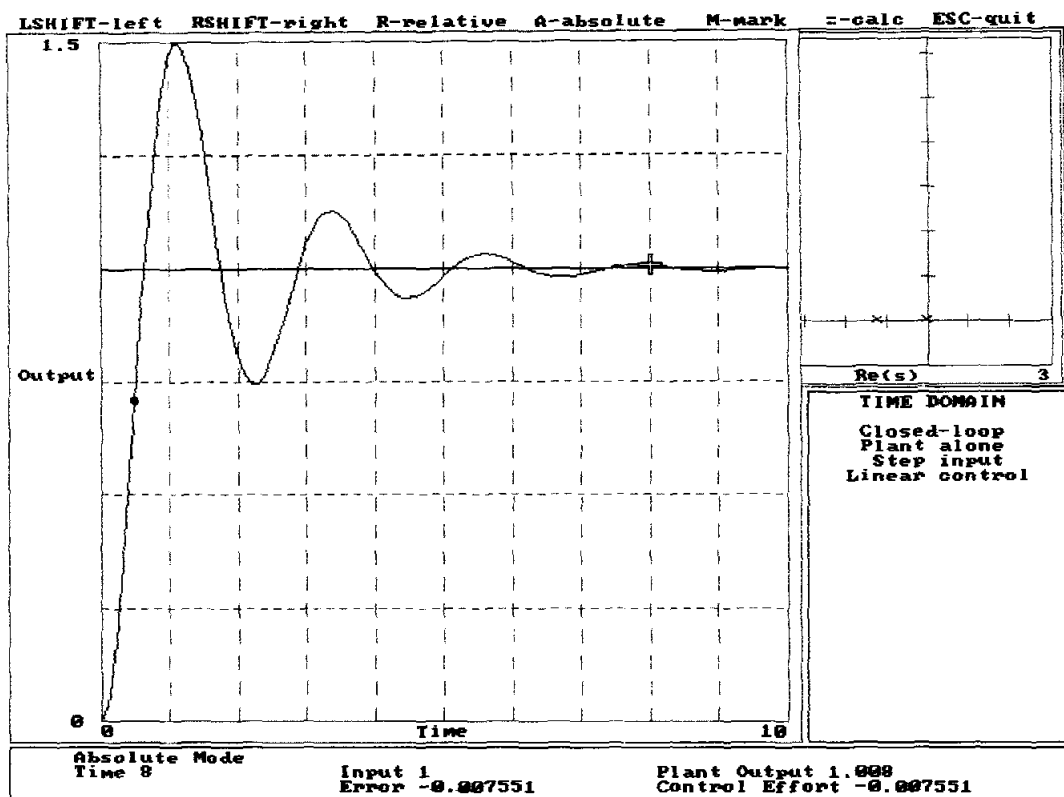
Grafik a.5. Tampilan layar Monitor PC saat motor berputar 2500 RPM



b. Simulasi Codas



Gambar b.1. Respon Transfer Function Motor Servo DC tanpa Sistem Pengendali



Gambar b.2. Respon Transfer Function Motor Servo DC dengan Sistem Pengendali

Tabel Pengukuran Awal Motor Servo Dc

Duty Cycle	Teg. PWM	RPM (1)	RPM (2)	RPM (3)	Arus Ia(1)	Arus Ia (2)	Arus Ia (3)	RPM rata-rata	Arus Ia rata2	w	Kb	PI	J	
10%	1.2	182	190	175	0.34	0.33	0.34	182.3333333	0.336666667	19.0842222	0.0381818	0.2453178	0.012854	
20%	2.4	487	481	484	0.36	0.36	0.36	477.3333333	0.36	49.9608889	0.0379497	0.68256	0.013662	
30%	3.6	794	784	749	0.38	0.37	0.38	775.6666667	0.376666667	81.1884444	0.037847	1.1573711	0.014256	
40%	4.8	1038	1068	1031	0.4	0.4	0.39	1045.666667	0.396666667	109.446444	0.038783	1.6837178	0.015384	
50%	6	1358	1345	1327	0.42	0.41	0.41	1343.333333	0.413333333	140.602222	0.0385579	2.2408178	0.015937	
60%	7.2	1667	1642	1606	0.43	0.43	0.42	1638.333333	0.426666667	171.478889	0.0385043	2.8171378	0.016428	
70%	8.4	1931	1953	1925	0.45	0.45	0.44	1936.333333	0.446666667	202.669556	0.0383613	3.4726844	0.017135	
80%	9.6	2234	2226	2125	0.45	0.46	0.46	2195	0.458666667	229.743333	0.0390029	4.0920378	0.017811	
90%	10.8	2534	2508	2525	0.47	0.47	0.47	2522.333333	0.47	264.004222	0.0384161	4.76674	0.018056	
100%	12	2832	2825	2820	0.48	0.48	0.47	2825.666667	0.476666667	295.753111	0.038318	5.4019044	0.018265	
	Jumlah Teg.							Jumlah RPM	Jumlah Arus Ia	Jumlah w	Jumlah Kb			Jumlah J
	66							14942	4.16	1563.92933	0.3077906			0.159788
	rata-rata							rata-rata	rata-rata	rata-rata	rata-rata			rata-rata
	6.6							1494.2	0.416	156.392933	0.0307791			0.015979

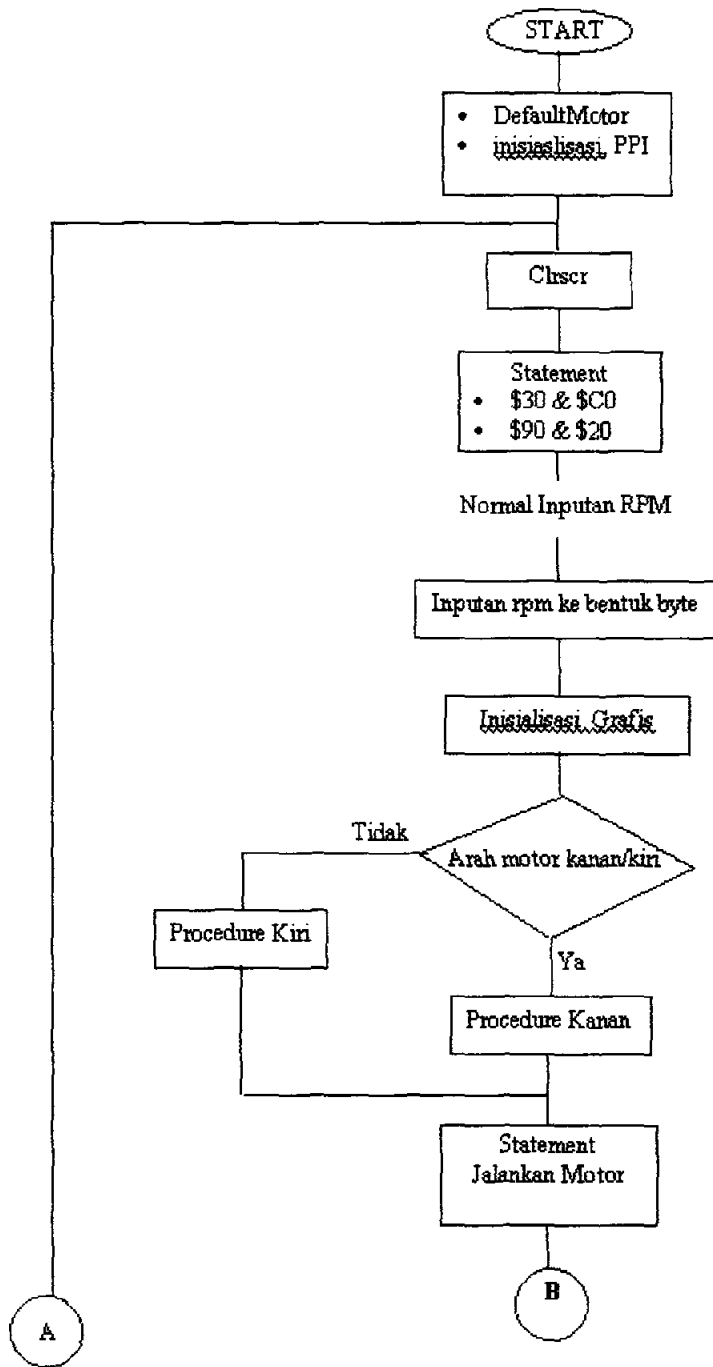
Kb = **0.030779** Vs/rad
 Kg = 0.030779 Nm/A
 J = 0.0159788 Kg/m2
 f = 0.0159788 Nms/rad

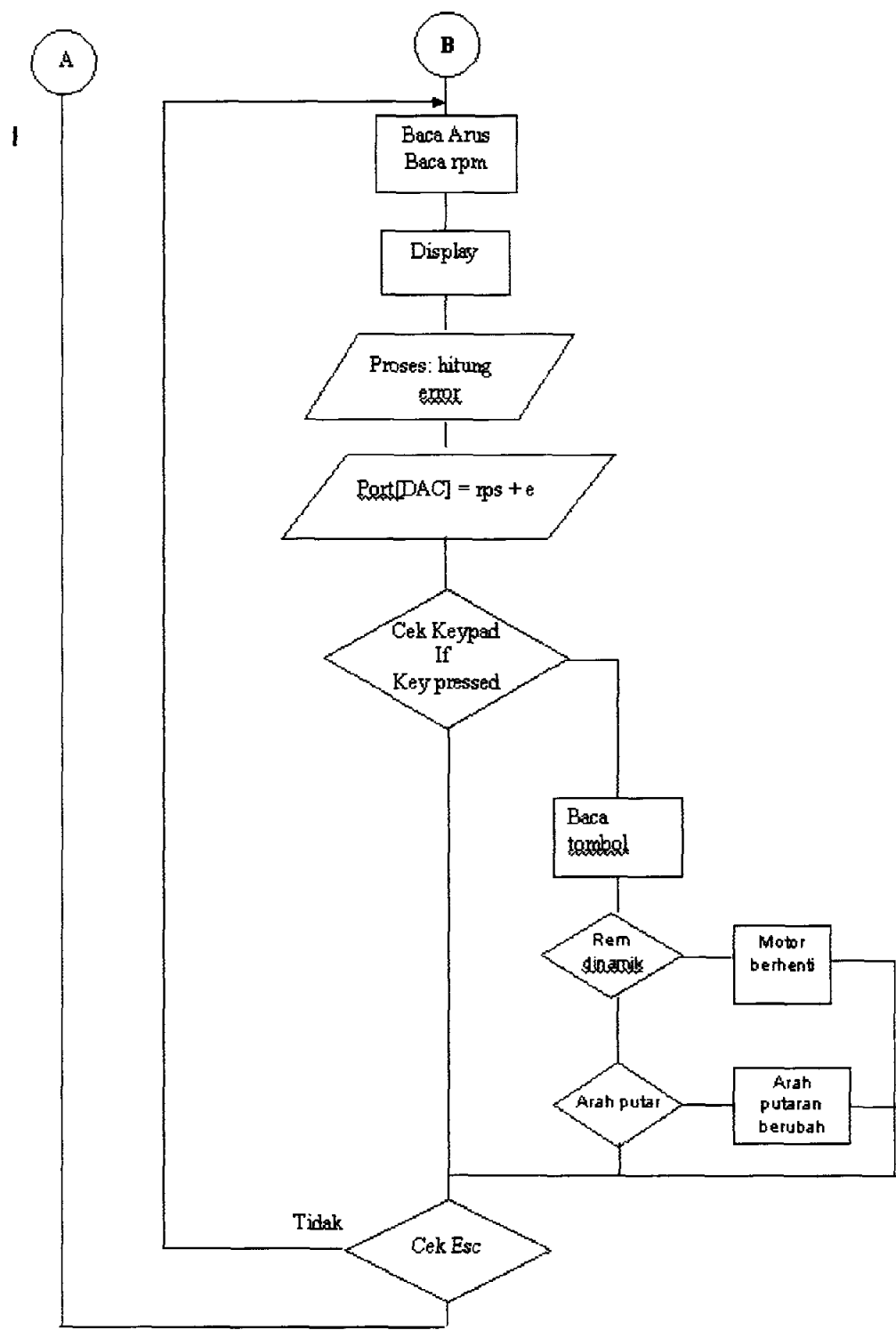
Tabel Pengukuran Sistem Pengaturan Motor Servo DC

Vpwm	Ra*la	dc rata2	Vpwm-ra*la	Kb	la ² *Ra	Vpwm*la	Pi	J	f=J
3.18	0.392	0.265	2.788	0.030372797	0.10976	0.8904	0.78064	0.0085044	0.008504383
4.0602	0.406	0.33835	3.6542	0.034278585	0.11774	1.177458	1.059718	0.0099408	0.00994079
4.1802	0.42	0.34835	3.7602	0.032526462	0.126	1.25406	1.12806	0.0097579	0.009757939
4.3398	0.434	0.36165	3.9058	0.030412845	0.13454	1.345338	1.210798	0.009428	0.009427982
4.404	0.434	0.367	3.97	0.029924999	0.13454	1.36524	1.2307	0.0092767	0.00927675
4.458	0.448	0.3715	4.01	0.026999367	0.14336	1.42656	1.2832	0.0086398	0.008639797
4.6782	0.448	0.38985	4.2302	0.026818795	0.14336	1.497024	1.353664	0.008582	0.008582014
4.8558	0.448	0.40465	4.4078	0.025947467	0.14336	1.553856	1.410496	0.0083032	0.008303189
5.04	0.462	0.42	4.578	0.025518584	0.15246	1.6632	1.51074	0.0084211	0.008421133
5.112	0.476	0.426	4.636	0.024283439	0.16184	1.73808	1.57624	0.0082564	0.008256369
5.28	0.476	0.44	4.804	0.023886593	0.16184	1.7952	1.63336	0.0081214	0.008121442
5.52	0.476	0.46	5.044	0.023803943	0.16184	1.8768	1.71496	0.0080933	0.008093341
5.6004	0.476	0.4667	5.1244	0.02292099	0.16184	1.904136	1.742296	0.0077931	0.007793137
5.6796	0.476	0.4733	5.2036	0.022269171	0.16184	1.931064	1.769224	0.0075715	0.007571518
5.9196	0.476	0.4933	5.4436	0.02243214	0.16184	2.012664	1.850824	0.0076269	0.007626928
6.24	0.476	0.52	5.764	0.022874377	0.16184	2.1216	1.95976	0.0077773	0.007777288
6.72	0.476	0.56	6.244	0.024069417	0.16184	2.2848	2.12296	0.0081836	0.008183602
7.0998	0.476	0.59165	6.6238	0.024490988	0.16184	2.413932	2.252092	0.0083269	0.008326936
V pwm Ttl				Kb Total				J total	f Total
92.368				0.473830961				0.1526045	0.152604538
Vpwm RT2				Kb Total RT2				J Rata2	f rata 2
5.1315				0.026323942				0.008478	0.00847803

Kb = 0.0263 Vs/rad
 Kg = 0.0263 Nm/A
 J = 0.0085 Kg/m²
 f = 0.0085 Nms/rad

LAMPIRAN FLOWCRAT :





LAMPIRAN PROGRAM

```
uses crt,graph;
const
    adc    = $300;
    dac    = $301;
    port_C = $302;
    port_CW = $303;
    CW     = $90;
    rem_dinamic_1 = $20;
    rem_dinamic_2 = $90;
    putar_kanan = $30;
    putar_kiri  = $C0;
    { inialisasi port PPI }
    { PPI aktif pada mode o}
    {set rem maju}
    {set rem mundur}
    {set arah putar motor}

var a_jangkar, V_tacho, arus : real;    {pemberian nilai variabel}
    ch,direct                : char;
    selesai,true,false      : boolean;
    e,status                  : byte;
    data_va,data_ia,putaran  : integer;
    tunda,gd,gm,x,y,x1      : integer;
    rpm2,rpm1                : byte;
    e1,arus1,rpm,arusl1     : real;
    int ,int2,putaranl1     : integer;

procedure fjdosij;
begin
    delay(1);
end;

procedure level_a;          { prosedur pembagian level RPM masukan}
begin
    rpm:=rpm - 110 ;
end;

procedure level_b;
begin
    rpm:=rpm - (180 + (0.01*rpm)) ;
end;

procedure level_c;
begin
    rpm:=rpm - (2 + (0.22*rpm));
end;

procedure level_d;
begin
    rpm:=rpm - ( 46 + (0.249*rpm)) ;
end;

procedure level_e;
begin
    rpm:=rpm - (30 + (0.29*rpm)) ;
end;
```

```
procedure level_f;  
begin  
rpm:=rpm - (110 + (0.28*rpm)) ;  
end;
```

```
procedure level_g;  
begin  
rpm:=rpm - (144 + (0.299*rpm)) ;  
end;
```

```
procedure level_h;  
begin  
rpm:=rpm - (142 + (0.3*rpm)) ;  
end;
```

```
procedure level_i;  
begin  
rpm:=rpm - (72 + (0.28*rpm)) ;  
end;
```

```
procedure level_j;  
begin  
rpm:=rpm - (20 + (0.26*rpm)) ;  
end;
```

```
procedure saluran_0;  
begin  
status := port[port_C];  
status := status AND $F8;  
port[port_c]:=status;  
status := status OR $00;  
port[port_C] := status;  
status:= status or $04;  
port[port_C] := status;  
delay(1);  
end;
```

{ prosedur untuk pembacaan data oleh
channel ADC0808 (multiplekser)}

```
procedure saluran_1;  
begin  
status := port[port_C];  
status := status AND $F8;  
port[port_c]:=status;  
status := status OR $01;  
port[port_C] := status;  
status:= status or $04;  
port[port_C] := status;  
delay(1);  
end;
```

```

procedure rem1;                                { prosedur pengaturan motor }
begin
  status := port[port_C];
  status := $0F AND status;
  status := rem_dinamic_1 OR status;
  port[port_C] := status;
end;

procedure rem2;
begin
  status := port[port_C];
  status := $0F AND status;
  status := rem_dinamic_2 OR status;
  port[port_C] := status;
end;

procedure kanan;
begin
  status := port[port_C];
  status := $0F AND status;
  status := putar_kanan OR status;
  port[port_C] := status;
end;

procedure kiri;
begin
  status := port[port_C];
  status := status and $0f;
  status := putar_kiri OR status;
  port[port_C] := status;
end;

procedure initgrafik;                          {prosedure tampilan grafis}
begin
  Gd:=Detect;
  InitGraph(Gd, Gm, ' ');
  if GraphResult <> grOk then
  Halt(1);
end;

function IntToStr(l: Longint): String;
{ Convert any integer type to a string }
var
  S: string[11];
begin
  Str(l, S);
  IntToStr := S;
end;

begin                                          {main program}
  selesai := false;
  port[port_cw] := cw;                        {default stop motor}
  port[dac] := $00;
  port[port_C] := $00;

```

```

repeat { baca ADC }
  clrscr;
  write(' Arah putaran motor : ');
  readln(direct);
  write(' Berapa RPM - nya : ');  readln(rpm);

      if(( rpm>800) and ( rpm<= 1050)) then          {penyesuai nilai input}
        begin
          level_a;
        end
      else
        if(( rpm>1050) and ( rpm<= 1250)) then
          begin
            level_b;
          end
        else
          if(( rpm>1250) and ( rpm<= 1450)) then
            begin
              level_c;
            end
          else
            if(( rpm>1450) and ( rpm<= 1650)) then
              begin
                level_d;
              end
            else
              if(( rpm>1650) and ( rpm<= 1850)) then
                begin
                  level_e;
                end
              else
                if(( rpm>1850) and ( rpm<= 2050)) then
                  begin
                    level_f;
                  end
                else
                  if(( rpm>2050) and ( rpm<= 2250)) then
                    begin
                      level_g;
                    end
                  else
                    if(( rpm>2250) and ( rpm<= 2450)) then
                      begin
                        level_h;
                      end
                    else
                      if(( rpm>2450) and ( rpm<= 2650)) then
                        begin
                          level_i;
                        end
                      else
                        if(( rpm>2650) and ( rpm<= 2850)) then

```

```

begin
level_j;
end

else
begin
port[dac]:=0;
end;

rpm2:=round(rpm*(255/2850));           {pengubah nilai desimal rpm ke heksa}
rpm1:=rpm2;
writeln(' ',rpm1);
initgrafik;

if upcase(direct)='R' then kanan       {inputan pilihan arah putar}
else kiri;
port[dac]:=rpm1;
{delay(500); }
ch:=#0;
x:=1;
x1:=1;
repeat
line(10,20,600,20);
saluran_1;
data_la := port[adc];
delay(5);
Arus_la := data_la;
arus1_la := arus;
saluran_2;
putaran_la := port[adc];           {rpm}
e:=round(abs(putaran*(255/220)-rpm2));           {penghitungan nilai error}
{delay(5); }
gotoxy(10,10);
writeln(' ',putaran*(84/110):4:2,' ',rpm1,' ');

if putaran<rpm2 then               {proses penyesuaian e=0}
begin
if (rpm1)>=255 then rpm1:=255
else
begin
rpm1:=rpm1+e;
port[dac]:=rpm1;
delay(3);
end;
{ port[dac]:=$ff;}
end
else
begin
if rpm1<=rpm2 then rpm1:=rpm2
else
begin
rpm1:=rpm1-e;
port[dac]:=rpm1;
delay(3);

```

```

        end;
    end;

    inc(x);
    if x=600 then
        begin
            x:=1;
            cleardevice;
            end;

    setviewport(1,1,600,240,true);
    moveto(x,150-round(putaran1)div 2);
    lineto(x,150-round(putaran)div 2);
    setviewport(1,200,600,400,true);
    moveto(x,150-round(arus1));
    lineto(x,150-round(arus1));
    arus1:=arus1;
    putaran1:=putaran;
    setcolor(black);

    outtextxy(40,10,'RPM      =');
    outtextxy(40,30,'Arus Jangkar =');
    outtextxy(200,10,inttostr(int));
    outtextxy(240,30,'exp-3');
    outtextxy(200,30,inttostr(int2));
    int:=round(putaran*(2850/220));
    int2:=round(arus*(255/255)*2.2);
    delay(10);
    setcolor(white);
    outtextxy(40,10,'RPM      =');
    outtextxy(40,30,'Arus Jangkar =');
    outtextxy(200,10,inttostr(int));
    outtextxy(240,30,'exp-2');
    outtextxy(200,30,inttostr(int2));

    if keypressed then
        begin
            ch:=readkey ;
            case ch of
                #80:rem1;
                #72:rem2;
                #75:kanan;
                #77:kiri;
            end;
        end;
    until ch=#27;
    closegraph;
    port[dac]:=0;
    writeln('Setting');
    ch:=readkey;
    until ch=#27;
    port[dac]:=0;
end.

```

{tampilan grafis dua layar}

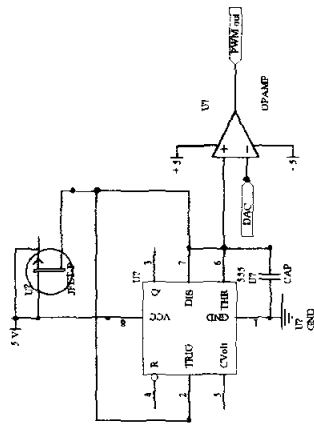
{tampilan nilai RPM}

{tampilan nilai arus}

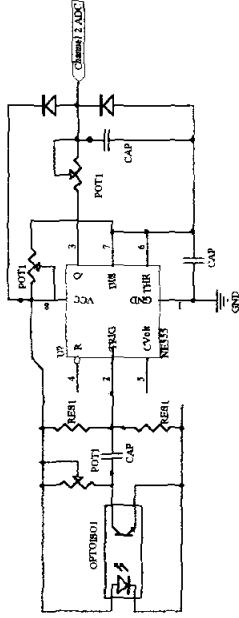
{pilihan pengaturan lanjut}

{kembali ke program utama}

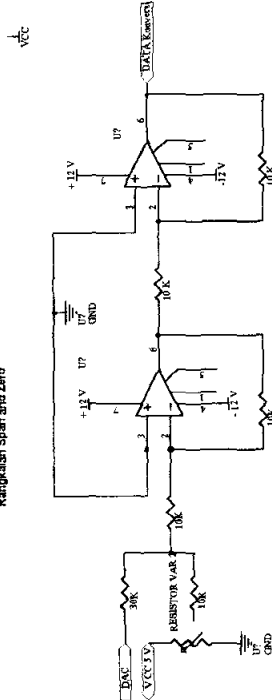
Rangkaian Pembangkit PWM



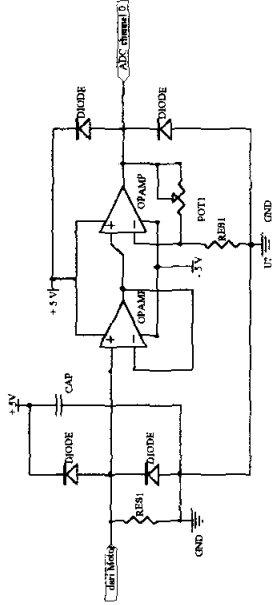
Rangkaian Pengukur Putaran



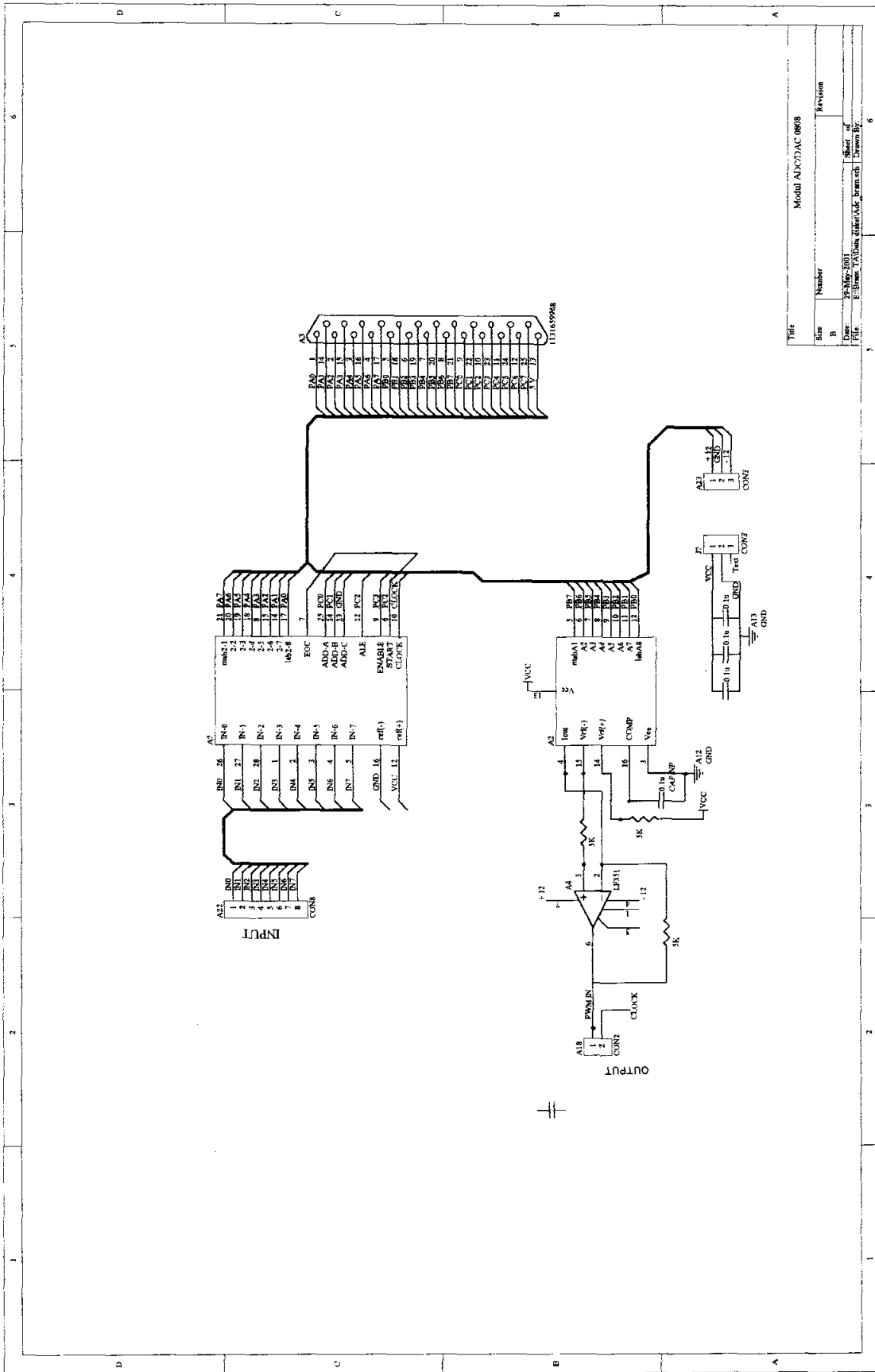
Rangkaian Span and Zero



Rangkaian Pengukur Arus

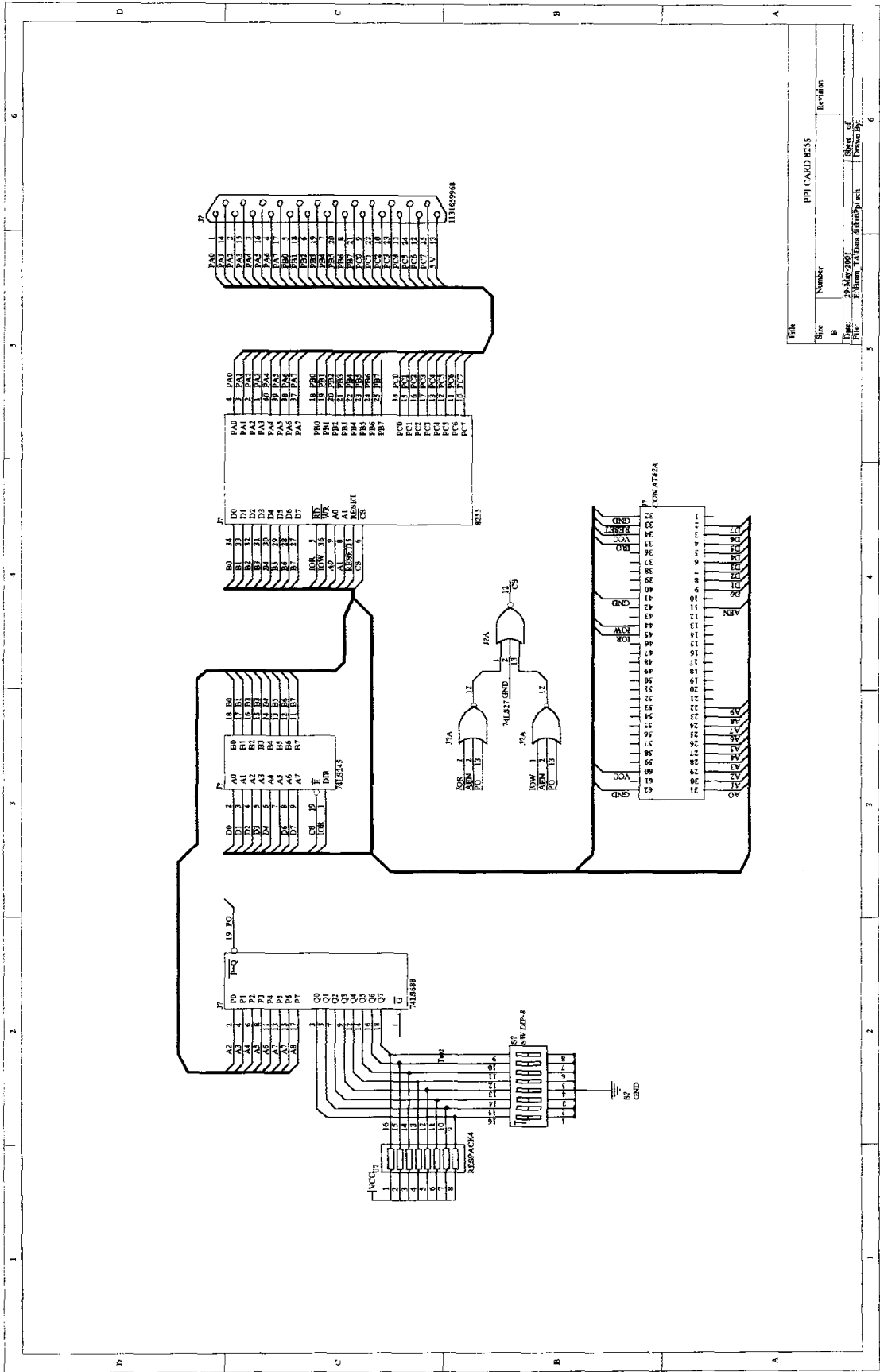


Title		Revision	
Size	Number		
B			
Date:	20 May 2008	Sheet of	6
File:	E:\Simulasi\Lab\Simulac	Drawn By:	



Pin	Signal
1	IN1
2	IN2
3	IN3
4	IN4
5	IN5
6	IN6
7	IN7
8	IN8
9	IN9
10	IN10
11	IN11
12	IN12
13	IN13
14	IN14
15	IN15
16	IN16
17	IN17
18	IN18
19	IN19
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93	IN93
94	IN94
95	IN95
96	IN96
97	IN97
98	IN98
99	IN99
100	IN100

Title		Modal A/D Converter (DAC) 08/68	
Size	Number	Sheet of	Revision
B		1	
Date:	29-May-2001	Drawn by:	
File:	E:\Bram\74133\Modal A/D Converter DAC	Checked by:	
		Drawn by:	



Title		PPI CARD 8255	
Size	Number	Revision	
B			
Drawn By	Checked By	Drawn By	Checked By
E. Urm, T. Adams			

MAXIM

Quad, High-Side MOSFET Drivers

MAX620/MAX621

General Description

The MAX620/MAX621 incorporate four MOSFET drivers and a charge-pump high-side power supply to power high-side switching and control circuits. The charge pump delivers a regulated output voltage 11V greater than VCC to the drivers, which then translate a TTL/CMOS input signal to a noninverted output that swings from ground to the high-side voltage. The outputs drive N-channel FETs in high-side or low-side switching applications, including a wide range of line- and battery-powered applications.

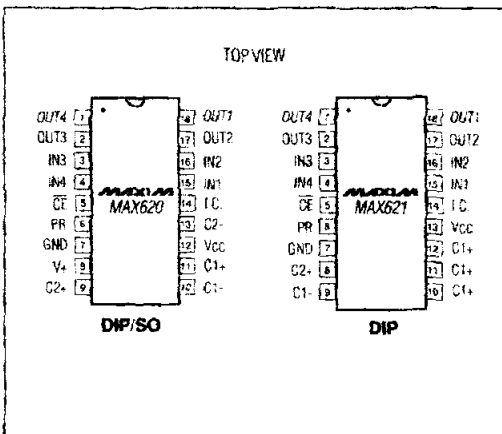
The MAX620/MAX621 are microprocessor compatible and feature undervoltage lockout capability. This lockout feature inhibits the FET driver outputs until the high-side voltage reaches the proper level, as indicated by a Power-Ready output.

The MAX620 requires three inexpensive charge-pump capacitors. The MAX621 has internal capacitors—no external components are needed.

Applications

- Portable Computer Battery Load Management
- High-Side Power, N-Channel MOSFET Switching
- Low-Side Switching from Low Supply Voltages
- Quad-Latching Level Translators
- H-Bridge Motor Drivers
- Stepper Motor Drivers

Pin Configurations



Features

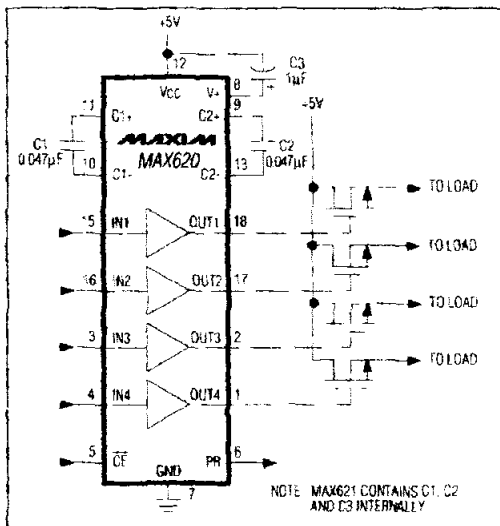
- ◆ Wide Operating Voltage Range
- ◆ Minimum Component Count
- ◆ Output Voltage Regulated to VCC Plus 11V (Typ)
- ◆ Low Quiescent Current – 70µA (Typ)
- ◆ Undervoltage Lockout
- ◆ Power-Ready Output
- ◆ Internal Quad Latch

Ordering Information

PART	TEMP. RANGE	PIN-PACKAGE
MAX620CPN	0°C to +70°C	18 Plastic DIP
MAX620CWN	0°C to +70°C	18 Wide SO
MAX620C/D	0°C to +70°C	Dice*
MAX620EPN	-40°C to +85°C	18 Plastic DIP
MAX620EWN	-40°C to +85°C	18 Wide SO
MAX621CPN	0°C to +70°C	18 Plastic DIP
MAX621EPN	-40°C to +85°C	18 Plastic DIP

*Contact factory for dice specifications.

Typical Operating Circuit



Quad, High-Side MOSFET Drivers

ABSOLUTE MAXIMUM RATINGS

V _{CC}	17V	Continuous Power Dissipation (T _A = +70°C)	
V+ to GND	30V	Plastic DIP (derate 8mW/°C above +70°C)	640mW
Inputs and Driver Outputs	(GND-0.3V) to (V+ + 0.3V)	Wide SO (derate 9.52mW/°C above +70°C)	762mW
PR Output	(GND-0.3V) to (V _{CC} + 0.3V)	Operating Temperature Ranges:	
Continuous Driver Output Current	25mA	MAX62 _ C	0°C to +70°C
V+ Output Current (MAX620 Only)	25mA	MAX62 _ E	-40°C to +85°C
		Storage Temperature Range	-65°C to +160°C
		Lead Temperature (Soldering, 10 sec)	+300°C

Stresses beyond those listed under 'Absolute Maximum Ratings' may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{CC} = +5V, T_A = T_{MIN} to T_{MAX}, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Supply Voltage	V _{CC}		4.5		16.5	V
High-Side Voltage (Note 1)	V+	I _{OUT} = 0, V _{CC} = 4.5V C1 = C2 = 0.047μF, C3 = 1μF	14.5	15.5	17.5	V
		I _{OUT} = 0, V _{CC} = 16.5V C1 = C2 = 0.01μF, C3 = 1μF (Note 2)	26.5	27.5	29.5	
		I _{OUT} = 250μA, V _{CC} = 5V, C1 = C2 = 0.047μF, C3 = 1μF	15	16	18	
		I _{OUT} = 500μA, V _{CC} = 16.5V, C1 = C2 = 0.01μF, C3 = 1μF (Note 2)	26.5	27.5	29.5	
Power-Ready Threshold	PRT	I _{OUT} = 100μA Sink (Notes 3, 4)	12.0	13.5	14.5	V
Power-Ready Output High	PR _{OH}	I _{SOURCE} = 100μA (Note 4)	3.6	4.7	5.0	V
Power-Ready Output Low	PR _{OL}	I _{SINK} = 1mA (Note 4)		0.1	0.4	V
Switching Frequency	f _o	I _{OUT} = 0, T _A = +25°C		70		KHz
Quiescent Supply Current	I _Q	MAX620 V _{CC} = 5V, C1 = C2 = 0.047μF, C3 = 1μF, T _A = +25°C, I _{OUT} = 0		70	500	μA
		MAX621 V _{CC} = 5V, T _A = +25°C, I _{OUT} = 0				
		MAX620 V _{CC} = 16.5V, C1 = C2 = 0.01μF, C3 = 1μF, T _A = +25°C, I _{OUT} = 0 (Note 5)		50	350	
		MAX621 V _{CC} = 16.5V, T _A = +25°C, I _{OUT} = 0				

Quad, High-Side MOSFET Drivers

MAX620/MAX621

ELECTRICAL CHARACTERISTICS (continued)

($V_{CC} = +5V$, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
HIGH-SIDE DRIVERS						
Input Threshold Low	V_{TL}				0.8	V
Input Threshold High	V_{TH}		2.4			V
Input Bias Current	I_B	$0V < V_{IN} < 5V$	-100		100	nA
Chip Enable Threshold Low	CE_{LO}				0.8	V
Chip Enable Threshold High	CE_{HI}		2.4			V
Minimum \overline{CE} Pulse Duration	T_{CE}		100	50		ns
Pull-Down Current	I_{CE}			10		μA
Data-Hold Time	T_{DH}			-10	10	ns
Data Set-Up Time	T_{SU}			50	100	ns
Data-Delay Time	T_{DD}	$V_{CE} = 0V, C_L = 120pF$		150		ns
Driver Output Rise Time	T_R	$C_L = 1000pF$		1.7		μs
Driver Output Fall Time	T_F	$C_L = 1000pF$		2.5		μs

Note 1: High-Side Voltage (V_+) is available only on the MAX620 and is measured with respect to GND. V_+ on the MAX621 is measured at an unloaded output. Capacitor values listed in the test conditions apply to the MAX620 only.

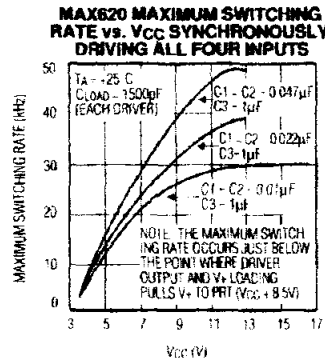
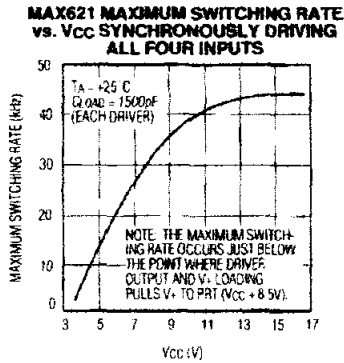
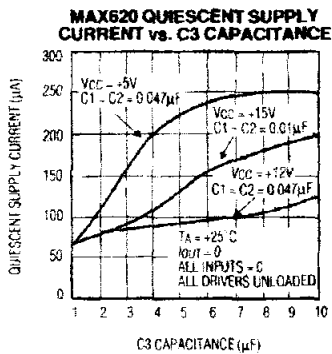
Note 2: For $V_{CC} > +13V$, on the MAX620 only, use $C1 = C2 = 0.01\mu F$, $C3 = 1\mu F$.

Note 3: Power-Ready Threshold is the voltage with respect to GND at V_+ when PR switches high ($PR_{OH} = V_{CC}$).

Note 4: For the MAX621, the Power-Ready levels are tested at wafer sort only.

Note 5: The MAX620 is tested for quiescent current at +16.5V using $C1 = C2 = 0.047\mu F$ to minimize test time. In normal operation above +13V, $C1$ and $C2$ must not exceed $0.01\mu F$.

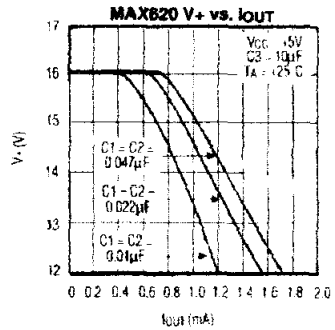
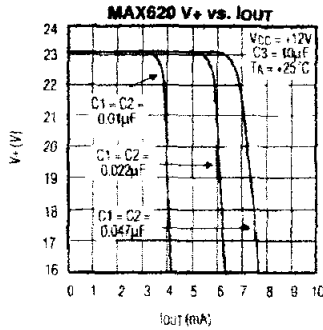
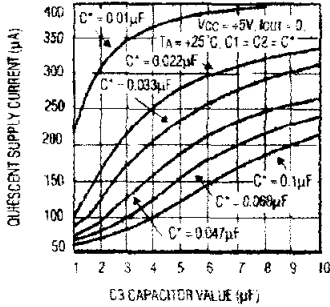
Typical Operating Characteristics



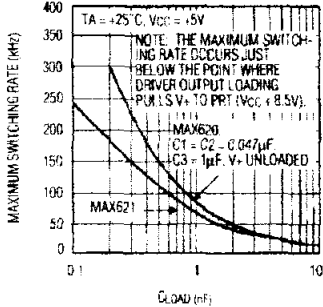
Quad, High-Side MOSFET Drivers

Typical Operating Characteristics (continued)

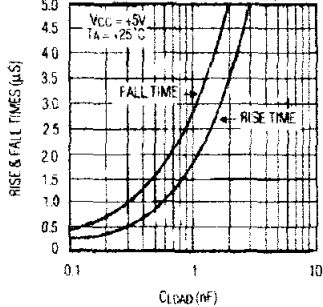
MAX620 QUIESCENT SUPPLY CURRENT vs. C3 CAPACITOR VALUE



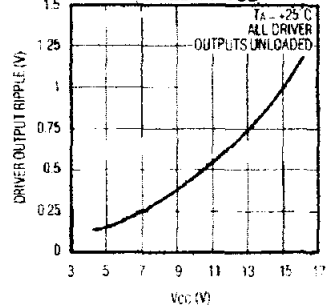
MAXIMUM SWITCHING RATE vs. LOAD SINGLE DRIVER



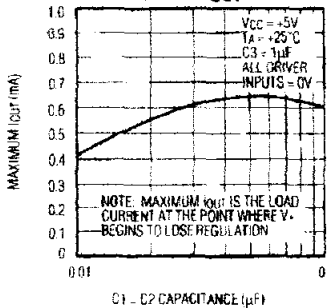
MAX620/MAX621 DRIVER RISE AND FALL TIME vs. LOAD



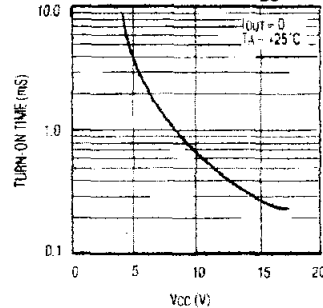
MAX621 DRIVER OUTPUT RIPPLE vs. VCC



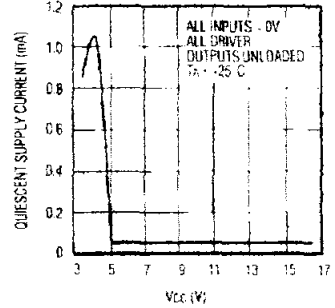
MAX620 MAXIMUM IOUT vs. C1 = C2



VCC TO POWER-READY HIGH DELAY vs. VCC



MAX621 QUIESCENT SUPPLY CURRENT vs. VCC

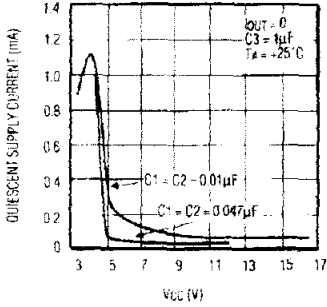


Quad, High-Side MOSFET Drivers

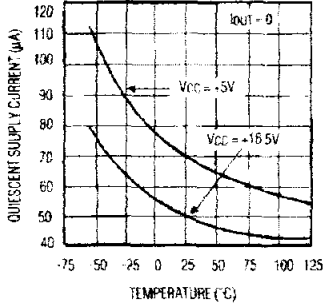
Typical Operating Characteristics (continued)

MAX620/MAX621

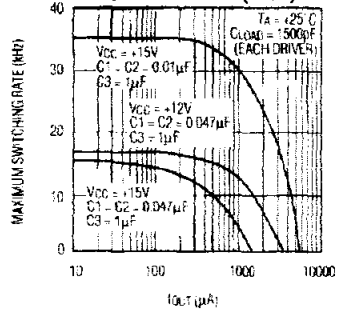
MAX620 QUIESCENT SUPPLY CURRENT vs. V_{CC}



MAX620/621 QUIESCENT SUPPLY CURRENT vs. TEMPERATURE

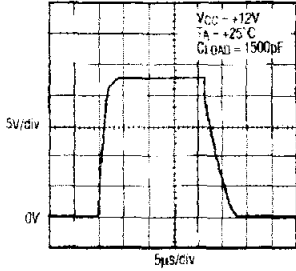


MAX620 MAXIMUM SWITCHING RATE vs. ADDITIONAL V₊ LOAD CURRENT (I_{OUT})

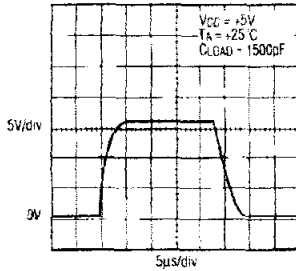


NOTE: THE MAXIMUM SWITCHING RATE OCCURS JUST BELOW THE POINT WHERE DRIVER OUTPUT AND V₊ LOADING PULLS V₊ TO PRT (V_{CC} + 0.5V).

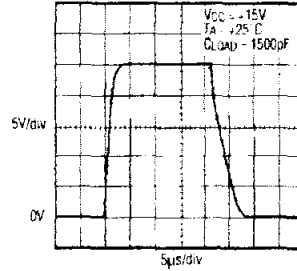
DRIVER OUTPUT SWITCHING WAVEFORM



DRIVER OUTPUT SWITCHING WAVEFORM



DRIVER OUTPUT SWITCHING WAVEFORM



Quad, High-Side MOSFET Drivers

MAX620/MAX621

Pin Description

PIN		NAME	FUNCTION
MAX620	MAX621		
1	1	OUT4	Driver Output 4
2	2	OUT3	Driver Output 3
3	3	IN3	TTL/CMOS Compatible Input to Driver 3. Connect to GND if unused.
4	4	IN4	TTL/CMOS Compatible Input to Driver 4. Connect to GND if unused.
5	5	CE	Chip Enable. Logic high inhibits input data. Logic low transfers input data to the quad latch and driver outputs. CE pulse must be at least 100ns. Connect to GND for direct data transfer to driver outputs.
6	6	PR	Power-Ready Output is a logic high equal to V _{CC} when V ₊ ≥ (V _{CC} plus 8.5V)
7	7	GND	Ground
8		V+	High-side voltage out. Equal to approximately V _{CC} plus 11V.
	8	C2+	Internally connected to secondary charge-pump capacitor. Make no connection to this pin.
9		C2+	Positive terminal to secondary charge-pump capacitor. Connect to 0.047μF capacitor. For V _{CC} > 13V, connect to 0.01μF.
	9	C1-	Internally connected to primary charge-pump capacitor. Make no connection to this pin.
10		C1-	Negative terminal to primary charge-pump capacitor. Connect to 0.047μF capacitor. For V _{CC} > 13V, connect to 0.01μF.
	10-12	C1+	Internally connected to primary charge-pump capacitor. Make no connection to these pins.
11		C1+	Positive terminal to primary charge-pump capacitor. Connect to 0.047μF capacitor. For V _{CC} > 13V, connect to 0.01μF.
12	13	V _{CC}	Supply Voltage. Connect to positive supply.
13		C2-	Negative terminal to secondary charge-pump capacitor. Connect to 0.047μF capacitor. For V _{CC} > 13V, connect to 0.01μF.
14	14	I.C.	Internal Connection. Make no connection to this pin.
15	15	IN1	TTL/CMOS Compatible Input to Driver 1. Connect to GND if unused.
16	16	IN2	TTL/CMOS Compatible Input to Driver 2. Connect to GND if unused.
17	17	OUT2	Driver Output 2
18	18	OUT1	Driver Output 1

DAC0808 8-Bit D/A Converter

General Description

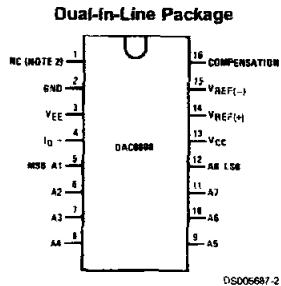
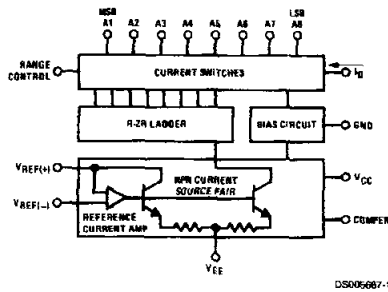
The DAC0808 is an 8-bit monolithic digital-to-analog converter (DAC) featuring a full scale output current settling time of 150 ns while dissipating only 33 mW with $\pm 5V$ supplies. No reference current (I_{REF}) trimming is required for most applications since the full scale output current is typically ± 1 LSB of $255 I_{REF}/256$. Relative accuracies of better than $\pm 0.19\%$ assure 8-bit monotonicity and linearity while zero level output current of less than $4 \mu A$ provides 8-bit zero accuracy for $I_{REF} \geq 2$ mA. The power supply currents of the DAC0808 is independent of bit codes, and exhibits essentially constant device characteristics over the entire supply voltage range.

The DAC0808 will interface directly with popular TTL, DTL or CMOS logic levels, and is a direct replacement for the MC1508/MC1408. For higher speed applications, see DAC0800 data sheet.

Features

- Relative accuracy: $\pm 0.19\%$ error maximum
- Full scale current match: ± 1 LSB typ
- Fast settling time: 150 ns typ
- Noninverting digital inputs are TTL and CMOS compatible
- High speed multiplying input slew rate: 8 mA/ μs
- Power supply voltage range: $\pm 4.5V$ to $\pm 18V$
- Low power consumption: 33 mW @ $\pm 5V$

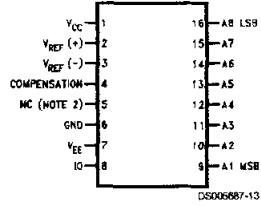
Block and Connection Diagrams



Top View
Order Number DAC0808
See NS Package M16A or N16A

Block and Connection Diagrams (Continued)

Small-Outline Package



Ordering Information

ACCURACY	OPERATING TEMPERATURE RANGE	N PACKAGE (N16A) (Note 1)		SO PACKAGE (M16A)
		DAC0808LCN	MC1408P8	DAC0808LCM
8-bit	$0^{\circ}\text{C} \leq T_A \leq +75^{\circ}\text{C}$	DAC0808LCN	MC1408P8	DAC0808LCM

Note 1: Devices may be ordered by using either order number.

Absolute Maximum Ratings (Note 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Power Supply Voltage

V_{CC}	+18 V_{DC}
V_{EE}	-18 V_{DC}
Digital Input Voltage, V5-V12	-10 V_{DC} to +18 V_{DC}
Applied Output Voltage, V_O	-11 V_{DC} to +18 V_{DC}
Reference Current, I_{14}	5 mA
Reference Amplifier Inputs, V14, V15	V_{CC}, V_{EE}
Power Dissipation (Note 4)	1000 mW
ESD Susceptibility (Note 5)	TBD

Storage Temperature Range	-65°C to +150°C
Lead Temp. (Soldering, 10 seconds)	
Dual-In-Line Package (Plastic)	260°C
Dual-In-Line Package (Ceramic)	300°C
Surface Mount Package	
Vapor Phase (60 seconds)	215°C
Infrared (15 seconds)	220°C

Operating Ratings

Temperature Range	$T_{MIN} \leq T_A \leq T_{MAX}$
DAC0808	$0 \leq T_A \leq +75^\circ C$

Electrical Characteristics

($V_{CC} = 5V$, $V_{EE} = -15V_{DC}$, $V_{REF}/R14 = 2mA$, and all digital inputs at high logic level unless otherwise noted.)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
E_r	Relative Accuracy (Error Relative to Full Scale I_O)	(Figure 4)				%
	DAC0808LC (LM1408-8)				± 0.19	%
	Settling Time to Within 1/2 LSB (Includes t_{PLH})	$T_A = 25^\circ C$ (Note 7), (Figure 5)		150		ns
t_{PLH}, t_{PHL}	Propagation Delay Time	$T_A = 25^\circ C$, (Figure 5)		30	100	ns
TCI_O	Output Full Scale Current Drift			± 20		ppm/°C
MSB	Digital Input Logic Levels	(Figure 3)				
V_{IH}	High Level, Logic "1"		2			V_{DC}
V_{IL}	Low Level, Logic "0"				0.8	V_{DC}
MSB	Digital Input Current	(Figure 3)				
	High Level	$V_{IH} = 5V$		0	0.040	mA
	Low Level	$V_{IL} = 0.8V$		-0.003	-0.8	mA
I_{15}	Reference Input Bias Current	(Figure 3)		-1	-3	μA
	Output Current Range	(Figure 3)				
		$V_{EE} = -5V$	0	2.0	2.1	mA
		$V_{EE} = -15V, T_A = 25^\circ C$	0	2.0	4.2	mA
I_O	Output Current	$V_{REF} = 2.000V$, $R14 = 1000\Omega$, (Figure 3)	1.9	1.99	2.1	mA
	Output Current, All Bits Low	(Figure 3)		0	4	μA
	Output Voltage Compliance (Note 3)	$E_r \leq 0.19\%$, $T_A = 25^\circ C$				
		$V_{EE} = -5V, I_{REF} = 1mA$			-0.55, +0.4	V_{DC}
		V_{EE} Below -10V			-5.0, +0.4	V_{DC}
SRI_{REF}	Reference Current Slew Rate	(Figure 6)	4	8		mA/ μs
	Output Current Power Supply Sensitivity	$-5V \leq V_{EE} \leq -16.5V$		0.05	2.7	$\mu A/V$
	Power Supply Current (All Bits Low)	(Figure 3)				
I_{CC}				2.3	22	mA
I_{EE}				-4.3	-13	mA
V_{CC}	Power Supply Voltage Range	$T_A = 25^\circ C$, (Figure 3)	4.5	5.0	5.5	V_{DC}
V_{EE}			-4.5	-15	-16.5	V_{DC}
	Power Dissipation					

Electrical Characteristics (Continued)

($V_{CC} = 5V$, $V_{EE} = -15V$, $V_{REF}/R14 = 2\text{ mA}$, and all digital inputs at high logic level unless otherwise noted.)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
	All Bits Low	$V_{CC} = 5V$, $V_{EE} = -5V$		33	170	mW
		$V_{CC} = 5V$, $V_{EE} = -15V$		106	305	mW
	All Bits High	$V_{CC} = 15V$, $V_{EE} = -5V$		90		mW
		$V_{CC} = 15V$, $V_{EE} = -15V$		160		mW

Note 2: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. DC and AC electrical specifications do not apply when operating the device beyond its specified operating conditions.

Note 3: Range control is not required.

Note 4: The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX} , θ_{JA} , and the ambient temperature, T_A . The maximum allowable power dissipation at any temperature is $P_D = (T_{JMAX} - T_A)/\theta_{JA}$ or the number given in the Absolute Maximum Ratings, whichever is lower. For this device, $T_{JMAX} = 125^\circ\text{C}$, and the typical junction-to-ambient thermal resistance of the dual-in-line J package when the board mounted is 100°C/W . For the dual-in-line N package, this number increases to 175°C/W and for the small outline M package this number is 100°C/W .

Note 5: Human body model, 100 pF discharged through a 1.5 k Ω resistor.

Note 6: All current switches are tested to guarantee at least 50% of rated current.

Note 7: All bits switched.

Note 8: Pin-out numbers for the DAL080X represent the dual-in-line package. The small outline package pinout differs from the dual-in-line package.

Typical Application

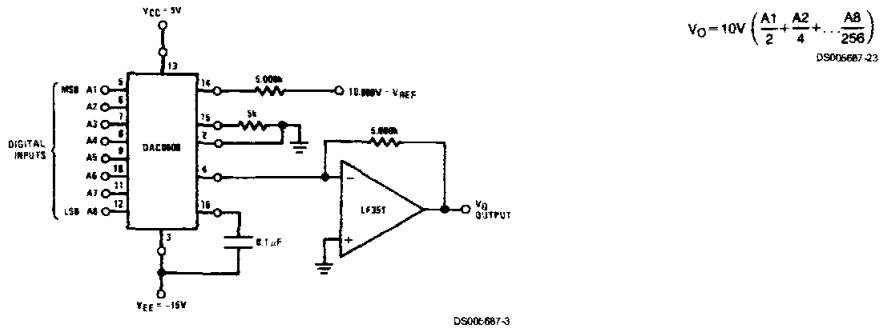
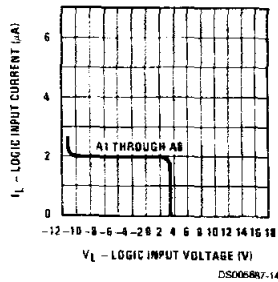


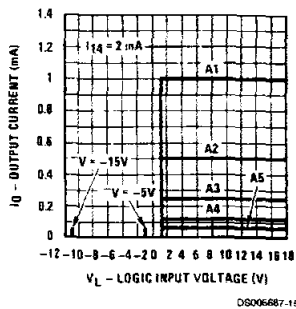
FIGURE 1. +10V Output Digital to Analog Converter (Note 8)

Typical Performance Characteristics $V_{CC} = 5V$, $V_{EE} = -15V$, $T_A = 25^\circ\text{C}$, unless otherwise noted

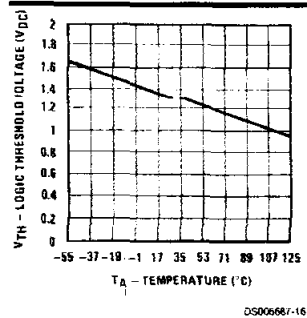
Logic Input Current vs Input Voltage



Bit Transfer Characteristics

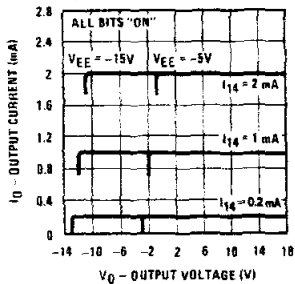


Logic Threshold Voltage vs Temperature



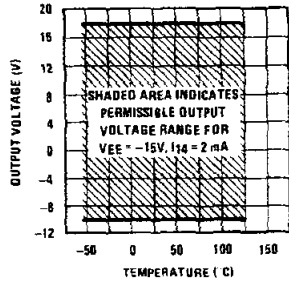
Typical Performance Characteristics $V_{CC} = 5V$, $V_{EE} = -15V$, $T_A = 25^\circ C$, unless otherwise noted (Continued)

Output Current vs Output Voltage (Output Voltage Compliance)



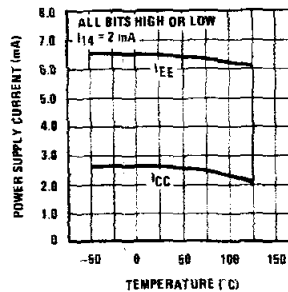
DS006687-17

Output Voltage Compliance vs Temperature



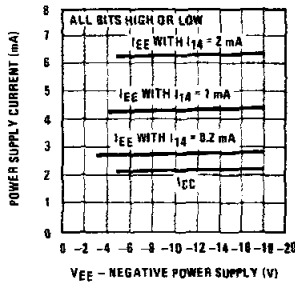
DS006687-18

Typical Power Supply Current vs Temperature



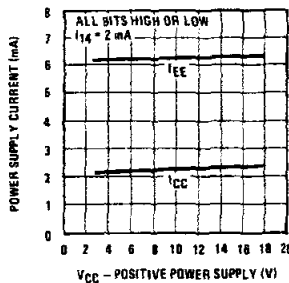
DS006687-19

Typical Power Supply Current vs VEE



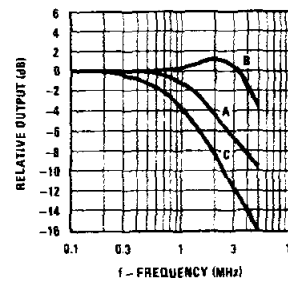
DS006687-20

Typical Power Supply Current vs VCC



DS006687-21

Reference Input Frequency Response



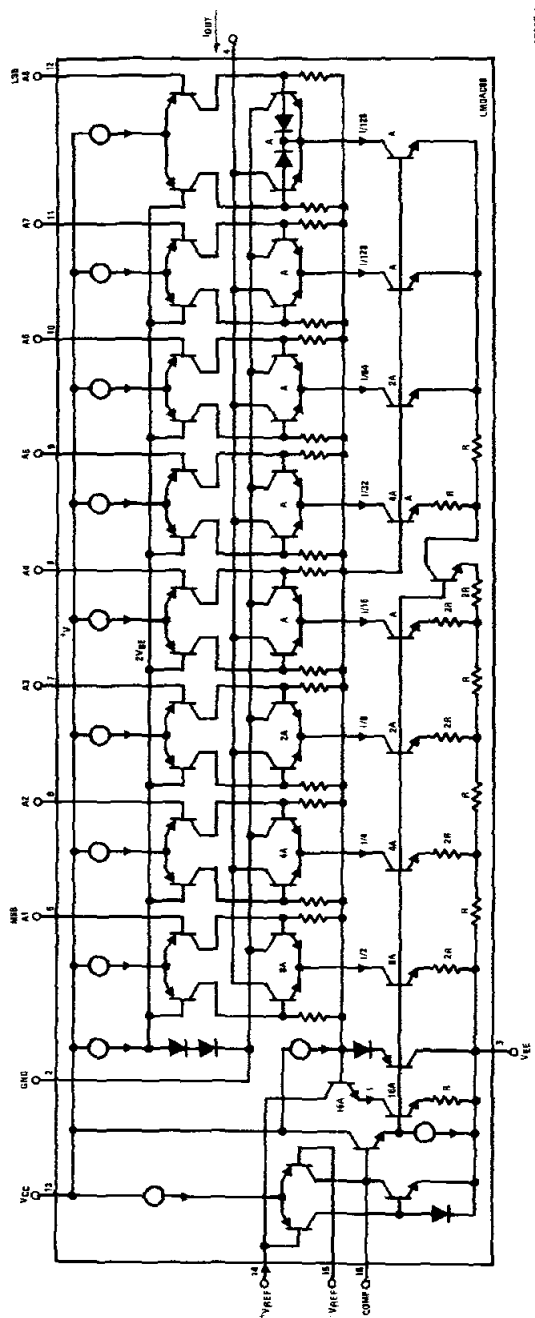
DS006687-22

Unless otherwise specified: $R_{14} = R_{15} = 1\text{ k}\Omega$, $C = 15\text{ pF}$, pin 16 to V_{EE} ; $R_L = 50\Omega$, pin 4 to ground.

Curve A: Large Signal Bandwidth Method of Figure 7, $V_{REF} = 2\text{ Vp-p}$ offset 1V above ground.

Curve B: Small Signal Bandwidth Method of Figure 7, $R_L = 250\Omega$, $V_{REF} = 50\text{ mVp-p}$ offset 200 mV above ground.

Curve C: Large and Small Signal Bandwidth Method of Figure 9 (no op amp, $R_L = 50\Omega$), $R_S = 50\Omega$, $V_{REF} = 2V$, $V_S = 100\text{ mVp-p}$ centered at 0V.



080008B7-4

FIGURE 2. Equivalent Circuit of the DAC0808 Series (Note 8)

ADC0808/ADC0809 8-Bit μ P Compatible A/D Converters with 8-Channel Multiplexer

General Description

The ADC0808, ADC0809 data acquisition component is a monolithic CMOS device with an 8-bit analog-to-digital converter, 8-channel multiplexer and microprocessor compatible control logic. The 8-bit A/D converter uses successive approximation as the conversion technique. The converter features a high impedance chopper stabilized comparator, a 256R voltage divider with analog switch tree and a successive approximation register. The 8-channel multiplexer can directly access any of 8 single-ended analog signals.

The device eliminates the need for external zero and full-scale adjustments. Easy interfacing to microprocessors is provided by the latched and decoded multiplexer address inputs and latched TTL TRI-STATE® outputs.

The design of the ADC0808, ADC0809 has been optimized by incorporating the most desirable aspects of several A/D conversion techniques. The ADC0808, ADC0809 offers high speed, high accuracy, minimal temperature dependence, excellent long-term accuracy and repeatability, and consumes minimal power. These features make this device ideally suited to applications from process and machine control to consumer and automotive applications. For 16-channel multiplexer with common output (sample/hold port) see ADC0815 data sheet. (See AN-247 for more information.)

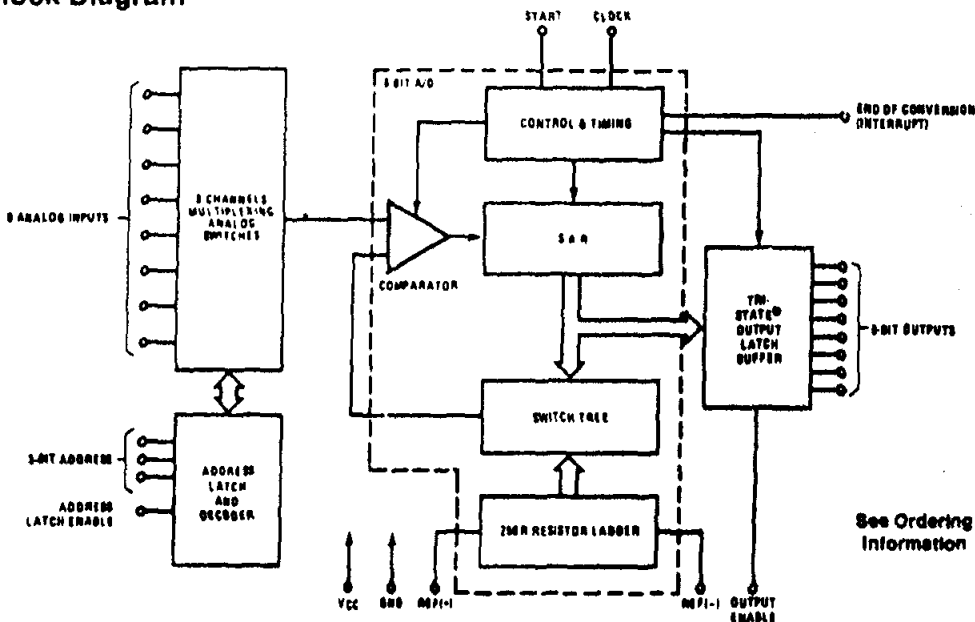
Features

- Easy interface to all microprocessors
- Operates ratiometrically or with 5 V_{CC} or analog span adjusted voltage reference
- No zero or full-scale adjust required
- 8-channel multiplexer with address logic
- 0V to 5V input range with single 5V power supply
- Outputs meet TTL voltage level specifications
- Standard hermetic or molded 28-pin DIP package
- 28-pin molded chip carrier package
- ADC0808 equivalent to MM74C949
- ADC0809 equivalent to MM74C949-1

Key Specifications

- | | |
|--------------------------|-------------------------------|
| ■ Resolution | 8 Bits |
| ■ Total Unadjusted Error | $\pm 1/2$ LSB and ± 1 LSB |
| ■ Single Supply | 5 V _{CC} |
| ■ Low Power | 15 mW |
| ■ Conversion Time | 100 μ s |

Block Diagram



See Ordering Information

TL/M/6672-1

ADC0808/ADC0809

Absolute Maximum Ratings (Notes 1 & 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Supply Voltage (V _{CC}) (Note 3)	6.5V
Voltage at Any Pin	-0.3V to (V _{CC} + 0.3V)
Except Control Inputs	
Voltage at Control Inputs	-0.3V to +15V
(START, OE, CLOCK, ALE, ADD A, ADD B, ADD C)	
Storage Temperature Range	-65°C to +150°C
Package Dissipation at T _A = 25°C	875 mW
Lead Temp. (Soldering, 10 seconds)	
Dual-In-Line Package (plastic)	260°C
Dual-In-Line Package (ceramic)	300°C
Molded Chip Carrier Package	
Vapor Phase (60 seconds)	215°C
Infrared (15 seconds)	220°C
ESD Susceptibility (Note 11)	400V

Operating Conditions (Notes 1 & 2)

Temperature Range (Note 1)	T _{MIN} ≤ T _A ≤ T _{MAX}
ADC0808CJ	-65°C ≤ T _A ≤ +125°C
ADC0808CCJ, ADC0808CCN,	
ADC0809CCN	-40°C ≤ T _A ≤ +85°C
ADC0808CCV, ADC0809CCV	-40°C ≤ T _A ≤ +85°C
Range of V _{CC} (Note 1)	4.5 V _{CC} to 6.0 V _{CC}

Electrical Characteristics

Converter Specifications: V_{CC} = 5 V_{CC} = V_{REF+}, V_{REF(-)} = GND, T_{MIN} ≤ T_A ≤ T_{MAX} and f_{CLK} = 640 kHz unless otherwise stated.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
V _{ERR}	ADC0808 Total Unadjusted Error (Note 5)	25°C T _{MIN} to T _{MAX}			± ½	LSB
					± ¾	LSB
	ADC0809 Total Unadjusted Error (Note 5)	0°C to 70°C T _{MIN} to T _{MAX}			± 1	LSB
					± 1½	LSB
	Input Resistance	From Ref(+) to Ref(-)	1.0	2.5		kΩ
	Analog Input Voltage Range	(Note 4) V(+) or V(-)	GND - 0.10		V _{CC} + 0.10	V _{CC}
V _{REF(+)}	Voltage, Top of Ladder	Measured at Ref(+)		V _{CC}	V _{CC} + 0.1	V
$\frac{V_{REF(+)} + V_{REF(-)}}{2}$	Voltage, Center of Ladder		V _{CC} /2 - 0.1	V _{CC} /2	V _{CC} /2 + 0.1	V
V _{REF(-)}	Voltage, Bottom of Ladder	Measured at Ref(-)	-0.1	0		V
I _{IN}	Comparator Input Current	f _c = 640 kHz, (Note 6)	-2	± 0.5	2	µA

Electrical Characteristics

Digital Levels and DC Specifications: ADC0808CJ 4.5V ≤ V_{CC} ≤ 5.5V, 55°C ≤ T_A ≤ +125°C unless otherwise noted
 ADC0808CCJ, ADC0808CCN, ADC0808CCV, ADC0809CCN and ADC0809CCV, 4.75V ≤ V_{CC} ≤ 5.25V, -40°C ≤ T_A ≤ +85°C unless otherwise noted

Symbol	Parameter	Conditions	Min	Typ	Max	Units
ANALOG MULTIPLEXER						
I _{OFF(+)}	OFF Channel Leakage Current	V _{CC} = 5V, V _{IN} = 5V, T _A = 25°C T _{MIN} to T _{MAX}		10	200 1.0	nA µA
I _{OFF(-)}	OFF Channel Leakage Current	V _{CC} = 5V, V _{IN} = 0, T _A = 25°C T _{MIN} to T _{MAX}	-200 -1.0	-10		nA µA

Electrical Characteristics (Continued)

Digital Levels and DC Specifications: ADC0808CJ, $4.5V \leq V_{CC} \leq 5.5V$, $-55^{\circ}C \leq T_A \leq +125^{\circ}C$ unless otherwise noted
 ADC0808CCJ, ADC0808CCN, ADC0808CCV, ADC0809CCN and ADC0809CCV, $4.75V \leq V_{CC} \leq 5.25V$, $-40^{\circ}C \leq T_A \leq +85^{\circ}C$ unless otherwise noted

Symbol	Parameter	Conditions	Min	Typ	Max	Units
CONTROL INPUTS						
$V_{IN(1)}$	Logical "1" Input Voltage		$V_{CC} - 1.5$			V
$V_{IN(0)}$	Logical "0" Input Voltage				1.5	V
$I_{IN(1)}$	Logical "1" Input Current (The Control Inputs)	$V_{IN} = 15V$			1.0	μA
$I_{IN(0)}$	Logical "0" Input Current (The Control Inputs)	$V_{IN} = 0$	-1.0			μA
I_{CC}	Supply Current	$f_{CLK} = 640 \text{ kHz}$		0.3	3.0	mA
DATA OUTPUTS AND EOC (INTERRUPT)						
$V_{OUT(1)}$	Logical "1" Output Voltage	$I_O = -360 \mu A$	$V_{CC} - 0.4$			V
$V_{OUT(0)}$	Logical "0" Output Voltage	$I_O = 1.8 \text{ mA}$			0.45	V
$V_{OUT(0)}$	Logical "0" Output Voltage EOC	$I_O = 1.2 \text{ mA}$			0.45	V
I_{OUT}	TRI-STATE Output Current	$V_O = 5V$ $V_O = 0$	3		3	μA

Electrical Characteristics

Timing Specifications $V_{CC} = V_{REF(+)} = 5V$, $V_{REF(-)} = GND$, $t_r = t_f = 20 \text{ ns}$ and $T_A = 25^{\circ}C$ unless otherwise noted.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
t_{WS}	Minimum Start Pulse Width	(Figure 5)		100	200	ns
t_{WALE}	Minimum ALE Pulse Width	(Figure 5)		100	200	ns
t_s	Minimum Address Set-Up Time	(Figure 5)		25	50	ns
t_H	Minimum Address Hold Time	(Figure 5)		25	50	ns
t_D	Analog MUX Delay Time From ALE	$R_S = 011$ (Figure 5)		1	2.5	μS
t_{H1}, t_{H0}	OE Control to Q Logic State	$C_L = 50 \text{ pF}$, $R_L = 10k$ (Figure 8)		125	250	ns
t_{1H}, t_{0H}	OE Control to HI-Z	$C_L = 10 \text{ pF}$, $R_L = 10k$ (Figure 8)		125	250	ns
t_c	Conversion Time	$f_c = 640 \text{ kHz}$, (Figure 5) (Note 7)	90	100	116	μS
f_c	Clock Frequency		.10	640	1260	kHz
t_{EOC}	EOC Delay Time	(Figure 5)	0		$8 + 2 \mu S$	Clock Periods
C_{IN}	Input Capacitance	At Control Inputs		10	15	pF
C_{OUT}	TRI-STATE Output Capacitance	At TRI-STATE Outputs, (Note 12)		10	15	pF

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. DC and AC electrical specifications do not apply when operating the device beyond its specified operating conditions.

Note 2: All voltages are measured with respect to GND, unless otherwise specified.

Note 3: A series diode exists, internally, from V_{CC} to GND and has a typical breakdown voltage of 7 V_{DC} .

Note 4: Two on-chip diodes are tied to each analog input which will forward conduct for analog input voltages one diode drop below ground or one diode drop greater than the V_{CC} supply. The spec allows 100 mV forward bias of either diode. This means that as long as the analog V_{IN} does not exceed the supply voltage by more than 100 mV, the output code will be correct. To achieve an absolute 0V_{DC} to 5V_{DC} input voltage range will therefore require a minimum supply voltage of 4.900 V_{CC} over temperature variations, initial tolerance and loading.

Note 5: Total unadjusted error includes offset, full-scale, linearity, and multiplexer errors. See Figure 3. None of these A/Ds requires a zero or full-scale adjust. However, if an all zero code is desired for an analog input other than 0.0V, or if a narrow full-scale span exists (for example: 0.5V to 4.5V full-scale) the reference voltages can be adjusted to achieve this. See Figure 13.

Note 6: Comparator input current is a bias current into or out of the chopper stabilized comparator. The bias current varies directly with clock frequency and has little temperature dependence (Figure 6). See paragraph 4.0.

Note 7: The outputs of the data register are updated one clock cycle before the rising edge of EOC.

Note 8: Human body model, 100 pF discharged through a 1.5 k Ω resistor.

Functional Description

Multiplexer. The device contains an 8-channel single-ended analog signal multiplexer. A particular input channel is selected by using the address decoder. Table I shows the input states for the address lines to select any channel. The address is latched into the decoder on the low-to-high transition of the address latch enable signal.

TABLE I

SELECTED ANALOG CHANNEL	ADDRESS LINE		
	C	B	A
IN0	L	L	L
IN1	L	L	H
IN2	L	H	L
IN3	L	H	H
IN4	H	L	L
IN5	H	L	H
IN6	H	H	L
IN7	H	H	H

CONVERTER CHARACTERISTICS

The Converter

The heart of this single chip data acquisition system is its 8-bit analog-to-digital converter. The converter is designed

to give fast, accurate, and repeatable conversions over a wide range of temperatures. The converter is partitioned into 3 major sections: the 256R ladder network, the successive approximation register, and the comparator. The converter's digital outputs are positive true.

The 256R ladder network (*Figure 1*) was chosen over the conventional $T/2R$ ladder because of its inherent monotonicity, which guarantees no missing digital codes. Monotonicity is particularly important in closed loop feedback control systems. A non-monotonic relationship can cause oscillations that will be catastrophic for the system. Additionally, the 256R network does not cause load variations on the reference voltage.

The bottom resistor and the top resistor of the ladder network in *Figure 1* are not the same value as the remainder of the network. The difference in these resistors causes the output characteristic to be symmetrical with the zero and full-scale points of the transfer curve. The first output transition occurs when the analog signal has reached $+ \frac{1}{4}$ LSB and succeeding output transitions occur every 1 LSB later up to full-scale.

The successive approximation register (SAR) performs 8 iterations to approximate the input voltage. For any SAR type converter, n-iterations are required for an n-bit converter. *Figure 2* shows a typical example of a 3-bit converter. In the ADC0808, ADC0809, the approximation technique is extended to 8 bits using the 256R network.

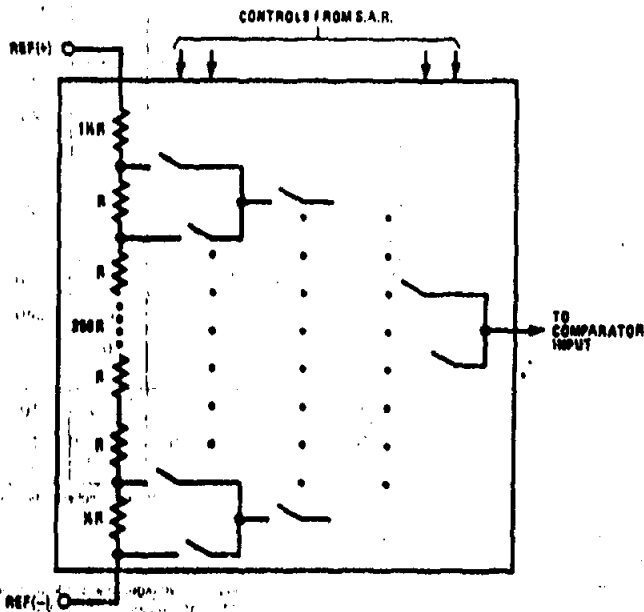


FIGURE 1. Resistor Ladder and Switch Tree

Functional Description (Continued)

The A/D converter's successive approximation register (SAR) is reset on the positive edge of the start conversion (SC) pulse. The conversion is begun on the falling edge of the start conversion pulse. A conversion in process will be interrupted by receipt of a new start conversion pulse. Continuous conversion may be accomplished by tying the end-of-conversion (EOC) output to the SC input. If used in this mode, an external start conversion pulse should be applied after power up. End-of-conversion will go low between 0 and 8 clock pulses after the rising edge of start conversion. The most important section of the A/D converter is the comparator. It is this section which is responsible for the ultimate accuracy of the entire converter. It is also the

comparator drift which has the greatest influence on the repeatability of the device. A chopper-stabilized comparator provides the most effective method of satisfying all the converter requirements.

The chopper-stabilized comparator converts the DC input signal into an AC signal. This signal is then fed through a high gain AC amplifier and has the DC level restored. This technique limits the drift component of the amplifier since the drift is a DC component which is not passed by the AC amplifier. This makes the entire A/D converter extremely insensitive to temperature, long term drift and input offset errors.

Figure 4 shows a typical error curve for the ADC0808 as measured using the procedures outlined in AN-179.

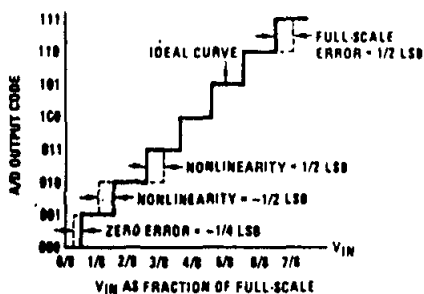


FIGURE 2. 3-Bit A/D Transfer Curve

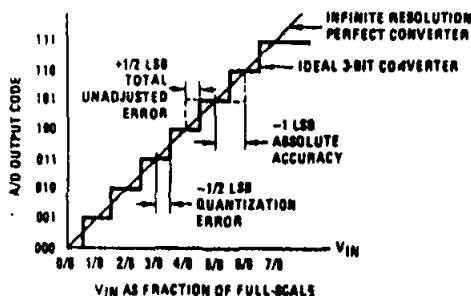


FIGURE 3. 3-Bit A/D Absolute Accuracy Curve

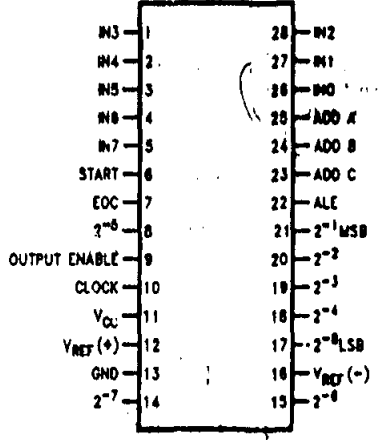


FIGURE 4. Typical Error Curve

TL/H/5072-3

Connection Diagrams

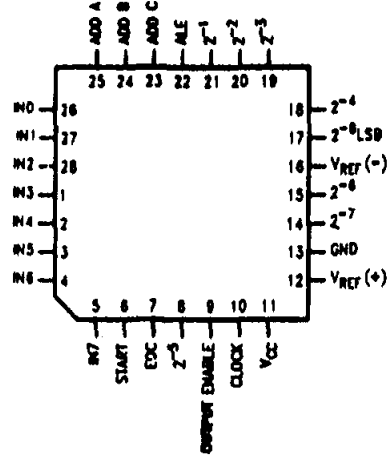
Dual-In-Line Package



Order Number ADC0808CCN, ADC0809CCN,
ADC0809CCJ or ADC0809CJ
See NS Package J28A or N28A

TL/H/0672-11

Molded Chip Carrier Package



Order Number ADC0808CCV or ADC0809CCV
See NS Package V28A

TL/H/0672-12

Timing Diagram

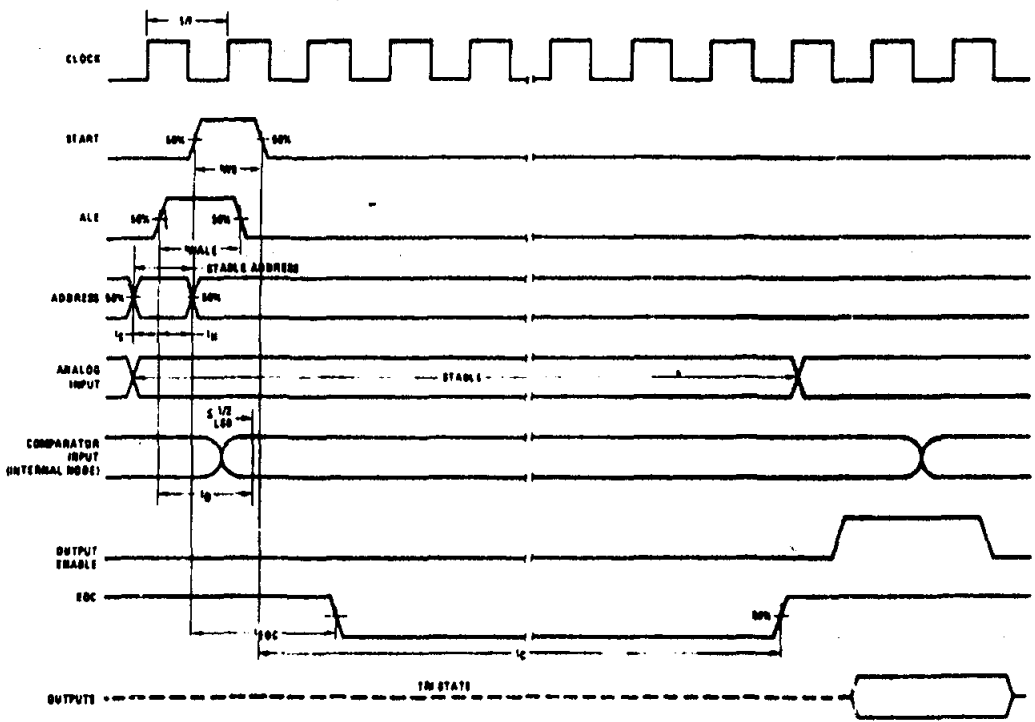



FIGURE 5

TL/H/0672-4

BIODATA



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